

A SIMPLE, HIGH QUALITY CD OUTPUT AMP

BY J.M. DIDDEN

I STARTED THE DESIGN for my CD output amp as a result of my ideas about a systems approach to audio design, on which I based my recent article (*TAA* 2/88). All components in a system should be simple, direct and designed with the rest of the chain in mind.

The first component I tackled was my CD player, specifically its analog output section. In *Fig. 1* the schematic for one channel is shown.

I don't think I must explain why I want to do without the output coupling electrolytic (2758) and the output series mute relay (1660, "kill"). I believe these types of components have been shown to audibly degrade the sound. If possible, I also would like to eliminate op amps 6675(a) and 6675(b). I believe this is potentially better than to try to find a better chip. Although we tend to think of op amps in terms of black

boxes, they contain tens of active devices which in one way or another all degrade the signal.

Refer to the Jung/Childress discussion on other pitfalls of op amps for this application.¹ As they explained, the first op amp provides current-to-voltage conversion and the first section of a third-order filter; the second provides a second-order low-pass filter. Relay 1658 (pre-emp) does the switching for the 50 μ sec de-emphasis. This relay is switched by the CD subcode information (*Fig. 6*).

Only a third-order output filter is necessary since my Philips CD 303 uses four-fold oversampling (176.4kHz instead of the basic 44.1kHz); more on this later.

In my effort to keep the circuitry as simple and direct as possible, I came up with the current-to-voltage converter of

Fig. 2. Actually, it can be done with just one transistor, but the DAC requires that its output remain within ± 10 mV from ground. The DAC output current is reflected in R_E and can be picked off at the emitter of Q2.

As you see, the output is referred to

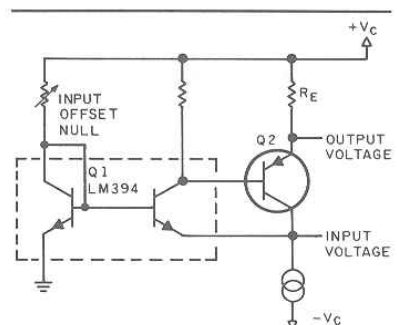


FIGURE 2: CD amp current-to-voltage converter.

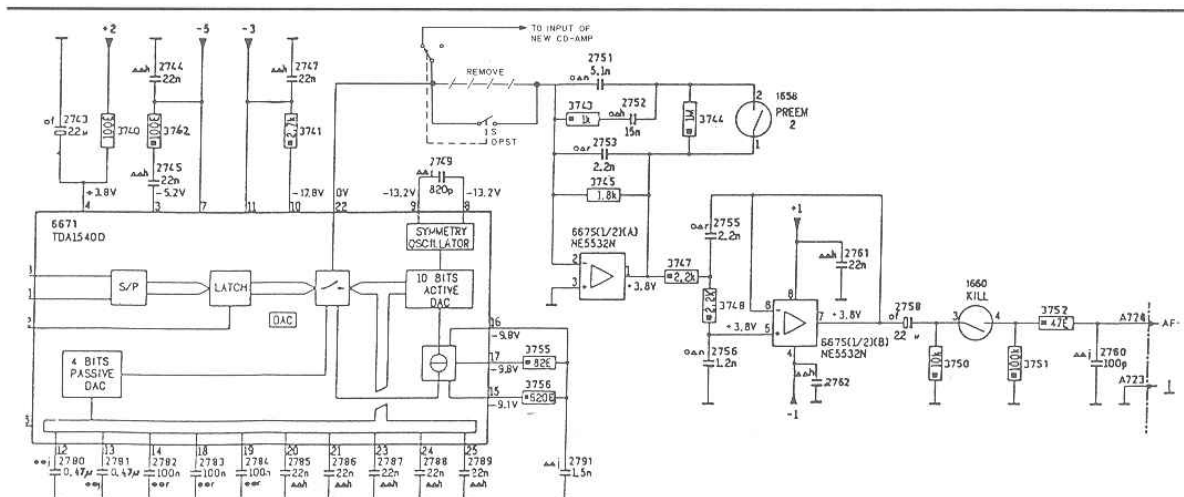


FIGURE 1: CD 303 analog output section (one channel). The author provides a switch to the input of the new CD amp.

frequency is switched is much more important, and while close tolerance will normally satisfy this criterion, it can also be achieved by simply matching the values of R1 and R2, R7 and R8, and ideally, C1 and C2 as well, but parity here is not quite so important.

The supply for the oscillator can be from a pair of 9V batteries, in which case the addition of the C3 and C4 is recommended. It can, however, be anything up to $\pm 18V$ —the maximum

permissible for this IC. Higher supply voltages mean a higher output, with a lower THD content.

So now, how do we use it? First, set the record level control to maximum, then connect the oscillator to both the line inputs via a "Y" cable. Operate the oscillator in the 800Hz mode and adjust the preset on the oscillator board to read zero VU on both channels. Normally, you will not need to alter this. If your main record control is ganged, some balance adjustment may be needed.

Now, using the record control, reduce the input level so the level meter reads $-20dB$ below zero VU. Next, run the test cassette in the record mode and monitor the replay output. If the calibration is initially close, the level of the recorded 800Hz will be almost the same as the input level— $20dB$. If it isn't, adjust the preset record trimmers so the replay level is as close as you can get it. By the way, you don't have to use the metering facility on the machine. If you want calibration "on the

Continued on page 53

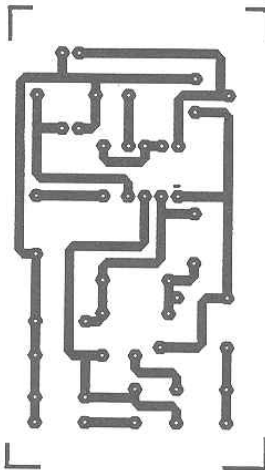


FIGURE 2: Circuit board pattern.

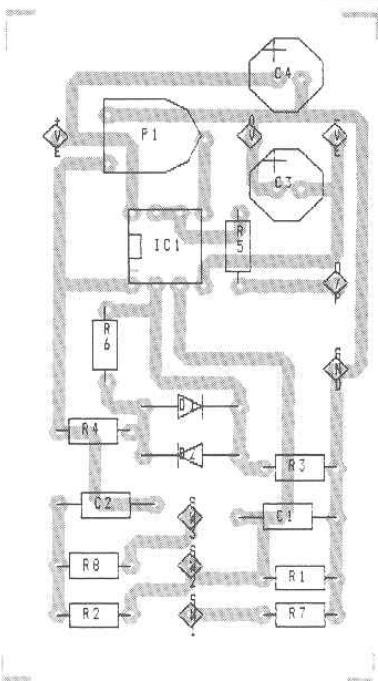


FIGURE 3: Component guide.

HI-FI CLASS A MOSFET MODULES

\$325 inc
P&P airmail

SPECIAL OFFER
SEE BELOW

50W to 250W Pure Class A

Build a top class audio dual monoblock power amplifier with an unparalleled level of performance, ideally suited for today's digital audio systems. Sage Audio's Supermos2 modules, designed by Les Sage, feature the most advanced and sonically accurate circuitry available today. Thousands have been sold world-wide and we are now the UK's largest exporter of pure Class A modular hi-fi kits.

FEATURES

- Pure Class A circuitry throughout
- Super fast slew rate ($685V/\mu\text{sec}$) for superbly detailed transients
- Top grade audiophile components
- Capacitor signature sound colorations eliminated
- Totally distortionless constant. IC, VCE circuitry, THD typically 1ppm
- Advanced PSU feed forward ripple elimination
- Localized feedback loops with extremely low overall feedback



- Max power output user defined from 50W-250W, dependent on PSU voltage

READY-BUILT AND TESTED: Complete with giant integral heatsink for cool running. Requires just a simple dual rail PSU and suitable case, full details supplied with the modules. Note these are ready-built monoblock modules and two are required for stereo.

TECHNICAL SPECS: THD, 0.0001%; IMD, 0; Slew rate, $685V/\mu\text{sec}$; f, $-3dB$, 0.5Hz-350kHz; Damping Factor, 940; O/P current capacity, 80A.

There's so much more that we couldn't possibly describe these modules fully in this ad, to receive an 8-page glossy brochure describing this and all our audio products send \$6 now, (sorry but no money no brochure).

SPECIAL OFFER FREE AIRMAIL—The first 100 inquiries requesting our brochure will receive a discount voucher worth \$50 off the price of two modules. Total cost includes free airmail just \$600 a pair.

SAGE AUDIO SAGE AUDIO, Construction House, Whitley Street, Bingley, Yorks. BD16 4JH. England.

Fast Reply #AD196

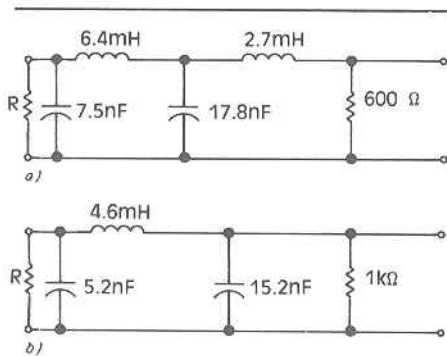


FIGURE 3: Passive output filter: (a) for two-times oversampling (fourth-order); (b) four four-times oversampling (third-order).

the positive supply line; I had to decide whether to use a high quality supply or add another stage. Since a high quality supply is desirable anyway, I chose not to introduce more active circuitry. The use of the LM394 provides an accurate and temperature-independent input offset null, which can be adjusted with the trimpot.

Output Filtering

All digital samples of the CD signal are coming off the disc at a rate of 44.1kHz, this is standard for any player. This means if we want to attenuate the

switching components, a standard 50dB in relation to the highest audio signal, we need a filter that rolls off at 50dB/octave. Such a filter is complex, expensive and ruins your phase response.

With four-fold oversampling, as used in my CD 303, each 44.1kHz sample is made into four samples that appear at the D/A converter at a rate of 176.4kHz. The process involves repeatedly multiplying each original sample. It is done entirely in the digital domain and thus has no influence on the quality.

So now our filter must roll off 50dB between 22kHz and 176.4kHz, or about

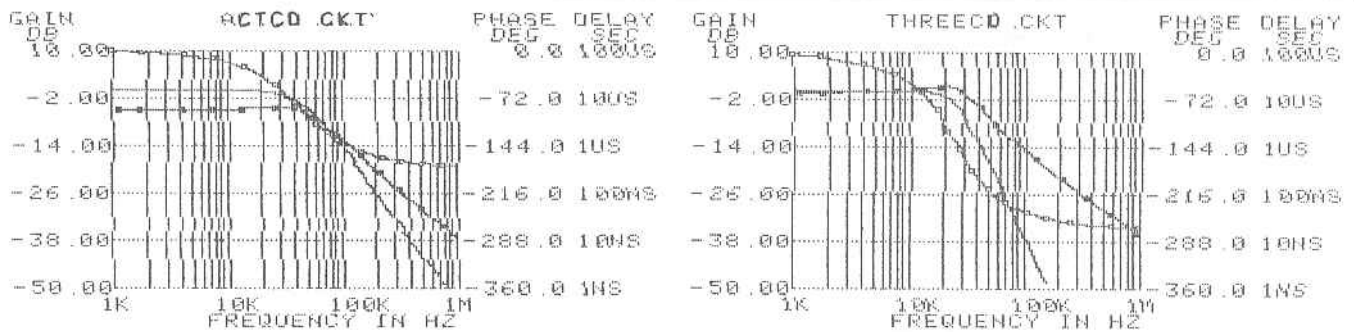


FIGURE 4: Simulated filter response curves: (a) original CD 303 second-order active filter (Fig. 1, IC 6675B); (b) third-order passive filter (Fig. 3b).

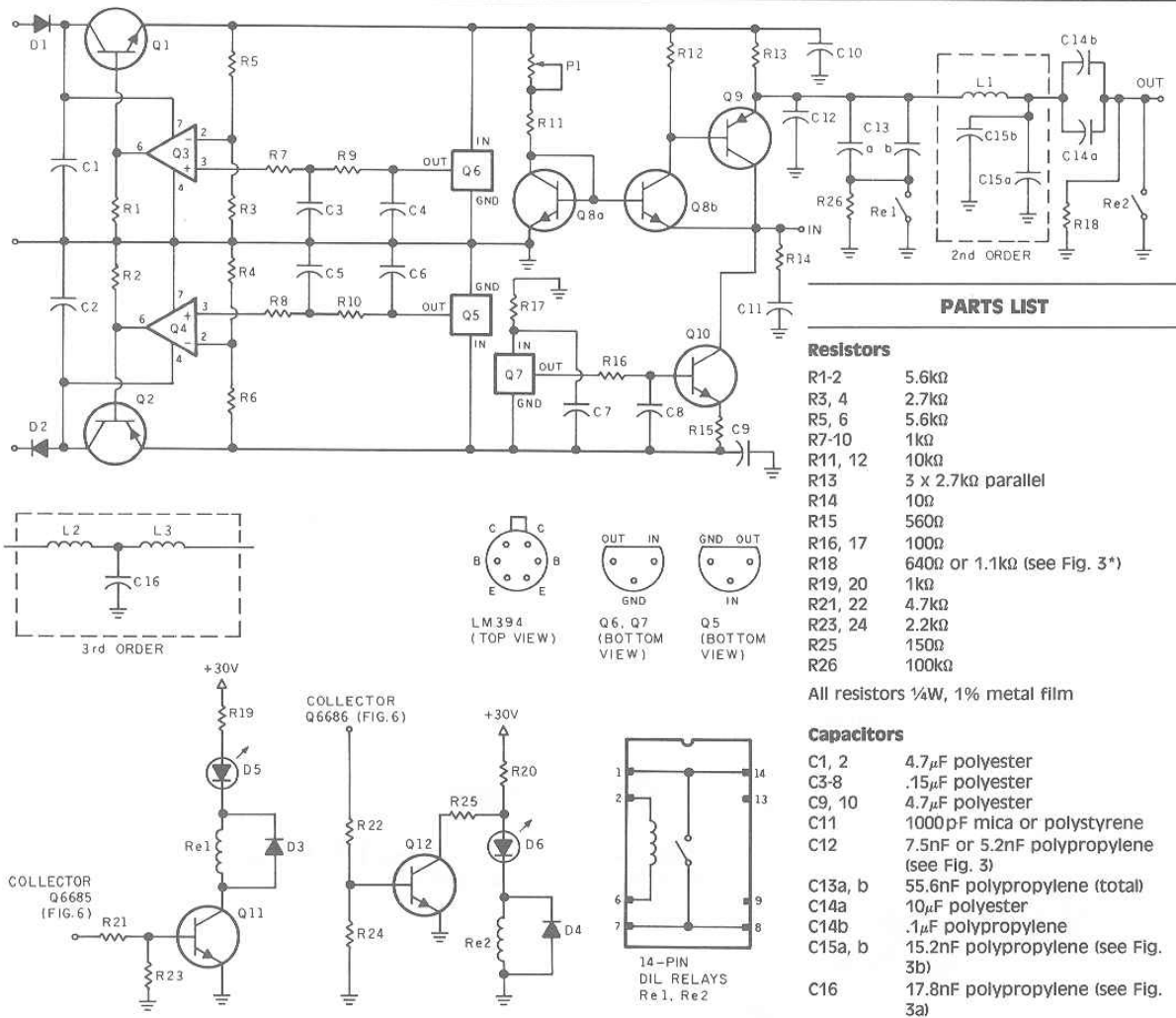


FIGURE 5: CD amp, complete schematic diagram.

18dB/octave. This can be done with the third-order filter, as shown in Fig. 1.

The active filter around 6675(b) with R3347, 3348 and C2755, 2756 provides 12dB/octave, another 6dB/octave is provided by R3745 and C2753 which also satisfies the default 5 μ sec de-emphasis.

The input rolloff of the preamp provides a safety margin where I normally use an RC filter at 30kHz. Some players use two-fold oversampling; the filter must roll off 50dB between 22kHz and 88.2kHz, or about 24dB/octave. My output amp can accommodate filters for both situations.

A passive filter is, I believe, inherently more neutral because of the lack of active circuitry in the signal path, provided you use high quality parts; but

the same requirement holds for the parts of an active filter.

Arthur Williams gives an excellent design "menu" for these classes of filters.² You can also use Robert Bullock's crossover design software³ which gives almost identical results. The filter circuits are shown in Fig. 3.

I generated the filter transfer curves from Fig. 4 on my computer with an analog simulation program.⁴ The measured responses closely follow these curves.

The complete schematic is shown in Fig. 5. The signal must negotiate just two transistors and some wire, instead of the tens of transistors hidden in an op amp. These kinds of circuits, simple, with low feedback, and without

Continued on page 30

PARTS LIST

Resistors

R1-2	5.6k Ω
R3, 4	2.7k Ω
R5, 6	5.6k Ω
R7-10	1k Ω
R11, 12	10k Ω
R13	3 x 2.7k Ω parallel
R14	10 Ω
R15	560 Ω
R16, 17	100 Ω
R18	640 Ω or 1.1k Ω (see Fig. 3*)
R19, 20	1k Ω
R21, 22	4.7k Ω
R23, 24	2.2k Ω
R25	150 Ω
R26	100k Ω

All resistors 1/4W, 1% metal film

Capacitors

C1, 2	4.7 μ F polyester
C3-8	.15 μ F polyester
C9, 10	4.7 μ F polyester
C11	1000pF mica or polystyrene
C12	7.5nF or 5.2nF polypropylene (see Fig. 3)
C13a, b	55.6nF polypropylene (total)
C14a	10 μ F polyester
C14b	.1 μ F polypropylene
C15a, b	15.2nF polypropylene (see Fig. 3b)
C16	17.8nF polypropylene (see Fig. 3a)

Semiconductors

Q1	BD139
Q2	BD140
Q3, 4	NE5534(A)
Q5	LM79L08
Q6	LM78L08
Q7	LM78L12
Q8a, b	LM394
Q9	BC560C
Q10	BC550C
Q11, 12	BC546
D1, 2	1N4003
D3, 4	1N4148
D5	yellow LED
D6	red LED
L1	4.6mH (see Fig. 3 and text)
L2	6.4mH (*)
L3	2.7mH (*)

Miscellaneous

Re1, 2	14-pin DIL 12V-reed relays
P1	multiturn trimpot (5k Ω)

* Assuming a 10k Ω volume control input at the preamp.

switches to +5V. The mute relay is activated by the collector of Q6686 going to zero volts.

I changed the topology for the mute switch. It is not in series with the signal, but is used to short the output signal. I believe this setup further decreases signal contamination.

The design also includes a wideband power supply for each channel.² The circuit needs $\pm 30V$ DC power. I did listening tests with four separate supplies for two channels, and with just two supplies with a common ground for both channels, and I could hear no difference. To minimize connections and solder joints, I used a gold-plated board-mounted RCA socket for the input. The output cable is directly connected to the circuit board, saving another solder and plug connection. The circuit pattern and parts placement for my unit are shown in Figs. 7 and 8, with the Parts List.

Building Coils

For the coils I used a type 2213 potcore (22 by 13mm), which is an industry standard size. They come in various "magnetic sizes," and we want an Al value of 400. (The Al value is the inductance (in millihenrys) for 1000 turns.)

The number of turns required $N = 1000 \sqrt{\text{inductance}/Al}$. Thus, for L1, $N = \#107$; for L2, $N = \#127$; and for

L3, $N = \#82$ turns. I used a double strand of enamel-insulated .22mm wire (I guess that's AWG 31). [AWG 31 is .227mm in diameter; AWG 32 is .202mm.—Ed.]

Use Litz if you can get it. Do not solder the wire ends to the core assembly terminals; I connected them directly to the board to avoid another two solder joints. See Fig. 8 for more details.

That's all there is to it. Do not forget to jumper the L3 terminals if you use the third-order filter. Discard the adjustable center slug if it was provided with your cores.

Customizing

Although I use the amp for my Philips CD 303 player, it would be particularly suited to any of the other Philips/Magnavox types.¹

You can use it for any brand, on two conditions. First, the player must have DACs with a current output. See also further considerations discussed in TAA^{1,4,5}

If your unit has voltage-output DACs, see whether you can replace them. This eliminates another not-so-hot op amp from the signal path. Most DAC types have interchangeable versions with either current or voltage output.

Also, your player must use at least two-fold, preferably four-fold oversampling, otherwise you need a much steep-

er filter. With the help of Williams, such filters are perfectly feasible, but I have not tried them. Another area where you are on your own is the mute and de-emphasis switching. You must find a suitable control signal that can be used to switch the DIL relays. This is normally not difficult, but a schematic of the output section of your unit is essential.

I am willing to provide customization advice if you send me a schematic diagram of your player through *Audio Amateur*.

As shown in Fig. 1, I provided a switch to select the original output or the current output, but this is just for comparison purposes. The current output is taken directly from the board through a fixed cable, avoiding the extra connections and joints of an output RCA jack.

Let me conclude by saying that this change caused a marked improvement in my CD's playback quality. The difference with my tuner and record player is much more pronounced. Much greater detail in the soundfield is audible; I can distinguish individual instruments and voices much better.

I now also detect greater difference in sound quality between CDs, where with the stock player they all sounded more or less alike.

This is a simple, inexpensive but very effective upgrade. Next, I will try to improve the circuitry around the TDA1540 DAC, per Jung/Childress.

REFERENCES

1. Jung, Walt and Hampton Childress, "POOGE-4," TAA 1,2/88.
2. Williams, Arthur B., *Electronic Filter Design Handbook*, McGraw-Hill, ISBN 0-07-070430-9, paragraphs 3 and 9.4; table 2.12.
3. Old Colony Crossover CAD program, OCSL SBK-F1A.
4. Microcomputer Circuit Analysis Program, Spectrum Software, 1021 S. Wolfe Rd., Sunnyvale, CA 94087, (408) 738-4387.
5. Didden, J.M., "A Wideband Power Supply," TAA 1/87, p. 22.
6. Childress, Hampton, "Modifying Yamaha's CD2 Player," TAA 3/86, p. 14.

The author resides overseas so if you wish to contact him and receive a reply, send your letter to Audio Amateur and enclose a large unstamped, self-addressed envelope and include two international postal coupons, available at your post office.

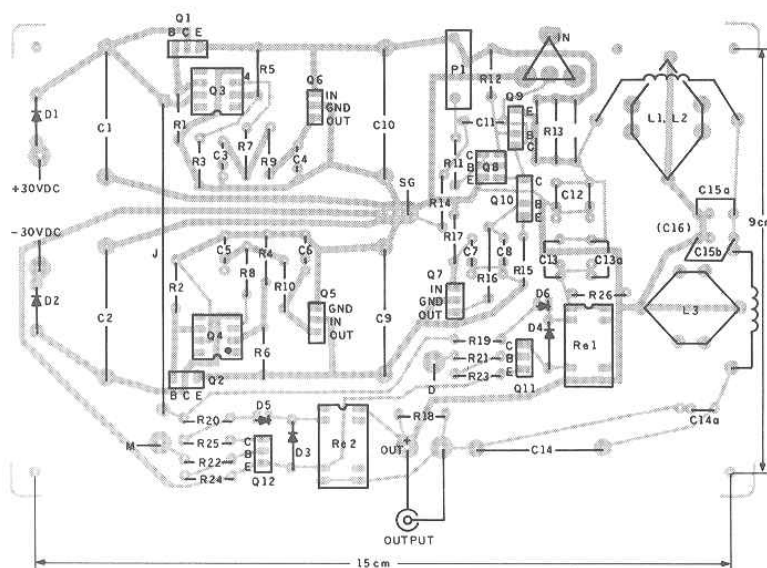


FIGURE 8: CD amp, parts placement guide. M is the mute control section; D is the 50µsec control (see Fig. 6).